

DEVELOPMENT OF SOLID-STATE INDUCTION MODULATORS FOR HIGH PRF ACCELERATORS

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Introduction

Researchers at the Lawrence Livermore National Laboratory (LLNL) and EG&G Energy Measurements are developing a new solid-state power system for two proposed accelerators. One of the accelerators is a circular arrangement of induction cells called a *recirculator*.¹ It is designed to accelerate heavy ions for an inertial fusion study that proposes to substitute heavy-ion beams for laser beams as the driver for fusion targets.² The other accelerator is a *linear induction accelerator* for electron beams called the Advanced Radiographic Machine (ARM). This accelerator generates multiple bursts of x rays for pulsed radiography by producing electron beams in rapid succession. Both accelerators require their induction cells to be pulsed at a very high repetition frequency (prf) for a short burst containing 5 to 15 pulses. The recirculator has a pulse schedule that varies in pulse width from 1 μ s to 400 ns and in prf from 50 to 150 kHz. The ARM accelerator has a pulse schedule that varies in pulse width from 1 μ s to 200 ns and in prf from 150 kHz to 1 MHz.

The need for complex pulse agility in these accelerators led us to examine solid-state switching components that have an on/off capability. The intrinsic speed of solid-state switching satisfies our high prf requirements, while the on/off switching action of some semiconductor devices enables us to select an arbitrary pulse width. To accommodate these requirements, we selected field effect transistors (FETs) as the preferred switching elements.

The same FET switching technology applies to both accelerators due to their similar pulse requirements. However, these two accelerators differ greatly in peak power and prf range. For example, the power system for the ARM accelerator must supply over 3 kA of beam-current loading to a 150-kV induction cell. Figure 1(a) shows the two induction-adder structures needed to achieve this high-power level. Each structure is composed of many small induction modulators that contribute 15 kV per modulator toward the total cell voltage. For contrast, Fig. 1(b) shows only an induction accelerating cell without an external power source, indicating that the recirculator power system is contained within each cell. The recirculator cell supplies only 5 kV to the ion beam per circulation and experiences very little beam-current loading.

For our research, two full-scale prototypes were built—a 5-kV induction recirculator cell and a single 15-kV induction modulator for the ARM accelerator. We discuss the general network features that are common to both machines, followed by performance and modeling data.

Circuits and Switches

The recirculator cell and the ARM induction modulator use the same basic circuit, shown in Fig. 2(a). Closing the switch S1 connects the energy storage capacitor C1 directly across the induction core L1. The voltage VC1 accelerates the beam, but it also drives a steadily increasing magnetization current, shown in Fig. 2(b). Once the acceleration pulse ends, S1 opens and the cell current begins to decrease. The sudden change in cell current results in a rapid voltage reversal on L1, which continues in the positive direction until the diode D1 is forward biased. The reset capacitor is precharged to the voltage VC2, which dictates the rate of cell current decay and helps to restore the magnetic flux density back to its original value. Additional details concerning the circuit and its operation may be found in earlier publications.^{3,4}

Report Documentation Page			Form Approved OMB No. 0704-0188		
<p>Public reporting burden for the collection of information is estimated to average 1 hour per response, including the time for reviewing instructions, searching existing data sources, gathering and maintaining the data needed, and completing and reviewing the collection of information. Send comments regarding this burden estimate or any other aspect of this collection of information, including suggestions for reducing this burden, to Washington Headquarters Services, Directorate for Information Operations and Reports, 1215 Jefferson Davis Highway, Suite 1204, Arlington VA 22202-4302. Respondents should be aware that notwithstanding any other provision of law, no person shall be subject to a penalty for failing to comply with a collection of information if it does not display a currently valid OMB control number.</p>					
1. REPORT DATE JUL 1995	2. REPORT TYPE N/A	3. DATES COVERED -			
4. TITLE AND SUBTITLE Development Of Solid-State Induction Modulators For High Prf Accelerators Of			5a. CONTRACT NUMBER		
			5b. GRANT NUMBER		
			5c. PROGRAM ELEMENT NUMBER		
6. AUTHOR(S)			5d. PROJECT NUMBER		
			5e. TASK NUMBER		
			5f. WORK UNIT NUMBER		
7. PERFORMING ORGANIZATION NAME(S) AND ADDRESS(ES) Lawrence Livermore National Laboratory, Livermore, CA, 94550			8. PERFORMING ORGANIZATION REPORT NUMBER		
9. SPONSORING/MONITORING AGENCY NAME(S) AND ADDRESS(ES)			10. SPONSOR/MONITOR'S ACRONYM(S)		
			11. SPONSOR/MONITOR'S REPORT NUMBER(S)		
12. DISTRIBUTION/AVAILABILITY STATEMENT Approved for public release, distribution unlimited					
13. SUPPLEMENTARY NOTES See also ADM002371. 2013 IEEE Pulsed Power Conference, Digest of Technical Papers 1976-2013, and Abstracts of the 2013 IEEE International Conference on Plasma Science. Held in San Francisco, CA on 16-21 June 2013. U.S. Government or Federal Purpose Rights License					
14. ABSTRACT					
15. SUBJECT TERMS					
16. SECURITY CLASSIFICATION OF: a. REPORT b. ABSTRACT c. THIS PAGE unclassified unclassified unclassified			17. LIMITATION OF ABSTRACT SAR	18. NUMBER OF PAGES 7	19a. NAME OF RESPONSIBLE PERSON

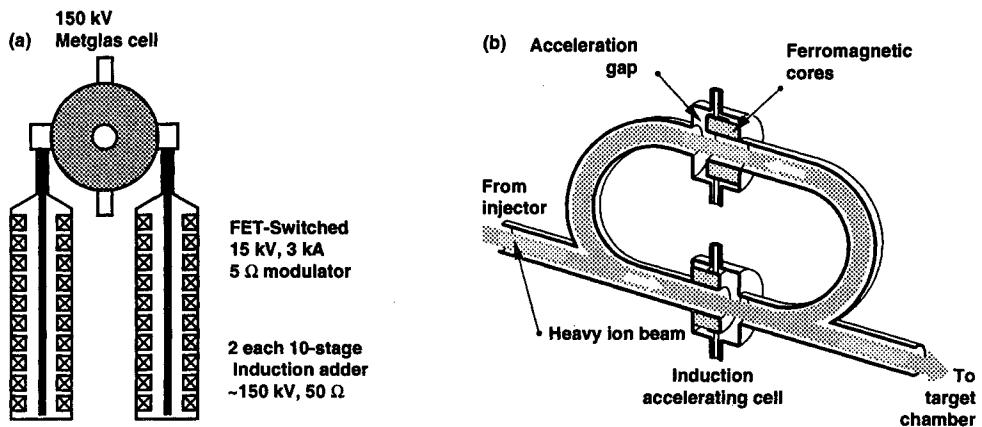


Figure 1. (a) ARM induction accelerator cell flanked by two induction-adder structures.
 (b) Illustration of a two-cell ion recirculator.

Table I lists the component values and maximum design voltages from Fig. 2(a) for both machines and provides design parameters as well.

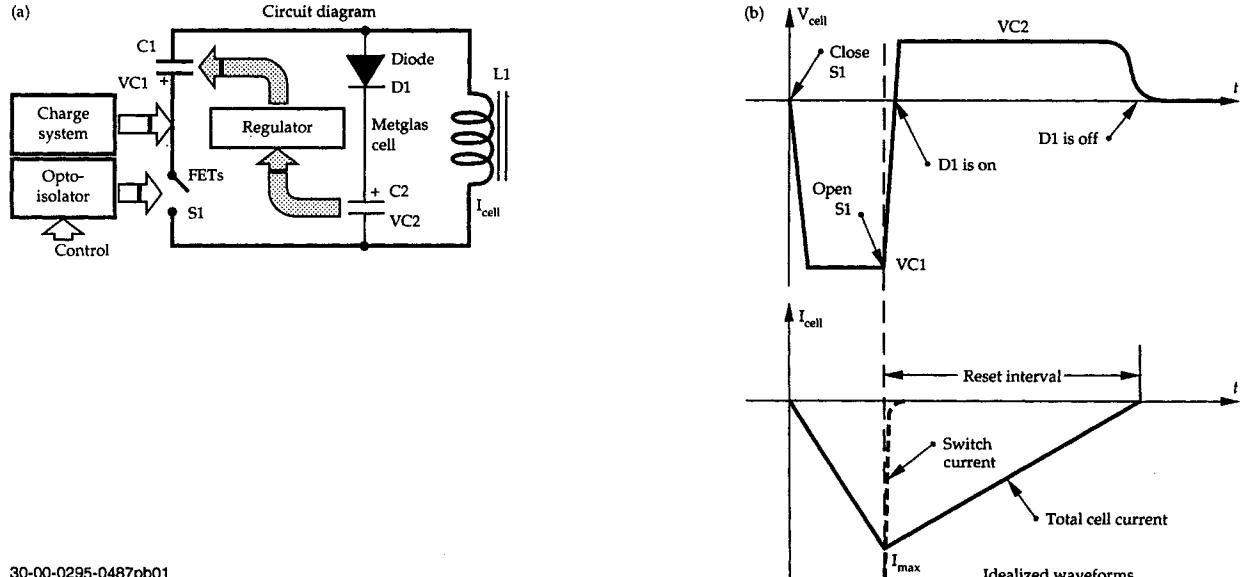
Performance Tests

The performance tests demonstrate the operating ranges of prf and pulse agility for both machines while examining several operational issues, including: opening and closing processes for the FETs, power delivery to external loads, and the system's stability during a burst.

Induction Recirculator Cell

Figure 3(a) shows the recirculator cell. The cylindrical assembly houses an induction core surrounded by capacitors (not visible) and is positioned on an insulated horizontal shaft. There are four switching modules attached to the core assembly and each module contains eight, 1-kV circuit cards. Figure 3(b) shows a rear view of one switching module.

Figure 4 compares single-pulse voltage and current data with simulations. The top row of simulated data was obtained using the Micro-Cap IV⁶ circuit analysis program for a 1-μs pulse with initial capacitor voltages of 5 kV and 900 V. The bottom row of experimental data shows the acceleration voltage



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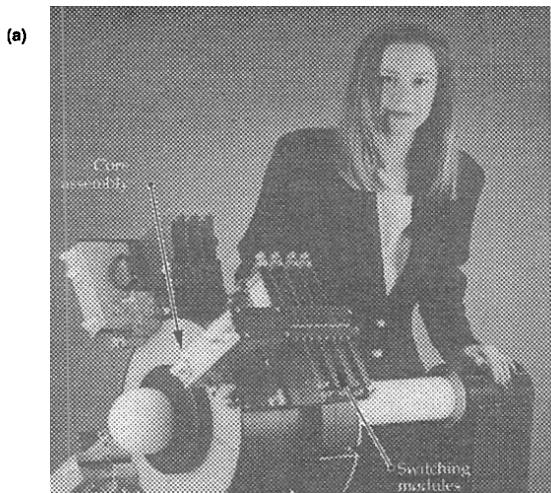
Figure 2. (a) Simplified circuit diagrams for the recirculator cell and ARM modulator. (b) Idealized current and voltage waveforms.

Table I. Comparison of component values and design parameters for the recirculator cell and ARM modulator.

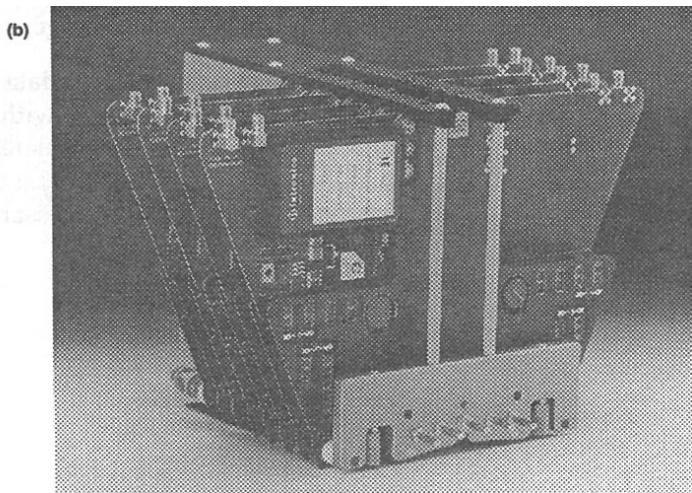
Components/Parameters	Recirculator Cell	ARM Modulator
Components and Voltages:		
C1	6 μ F	15.7 μ F
VC1	5 kV	15 kV
C2	136 μ F	15.7 μ F
VC2	900 V	5 kV
L1	18 μ H	38 μ H
Design Parameters:		
Max. current	800 A	3 kA
Pulse range	200 ns–1 μ s	200 ns–1 μ s
PRF range	250 kHz–150 kHz	1 MHz–150 kHz
Max. energy stored	75 J	1.8 kJ
No. of FETs	128	1,152
Metglas ⁵ core material	2605S-3A	2605S-3A
Machine o.d.	83.8 cm	94 cm
Approx. weight	80 kg	220 kg

detail and the switch current flowing through each module. Eight current signals are necessary to describe the total machine current, because each module is equipped with two current-viewing resistors (CVRs). The total switch current depicted in the simulation can be compared with a measured value by multiplying any of the four current signals by eight. The four measured currents, shown in Fig. 4, are a single CVR signal from each module, indicating an equal division of current.

The recirculator cell is designed for high prf burst operation and great pulse agility. The operational factors that define the prf limits within a burst are the capacitor charge voltages and the desired pulse width. An extreme test of prf performance is obtained with a combination of the minimum operating pulse width (200 ns) and the maximum accelerating and reset voltages (5 kV and 900 V). Figure 5(a)



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Figure 3. (a) Photograph illustrating the close integration between the switching modules and the induction core assembly. (b) Photograph of a single switching module.

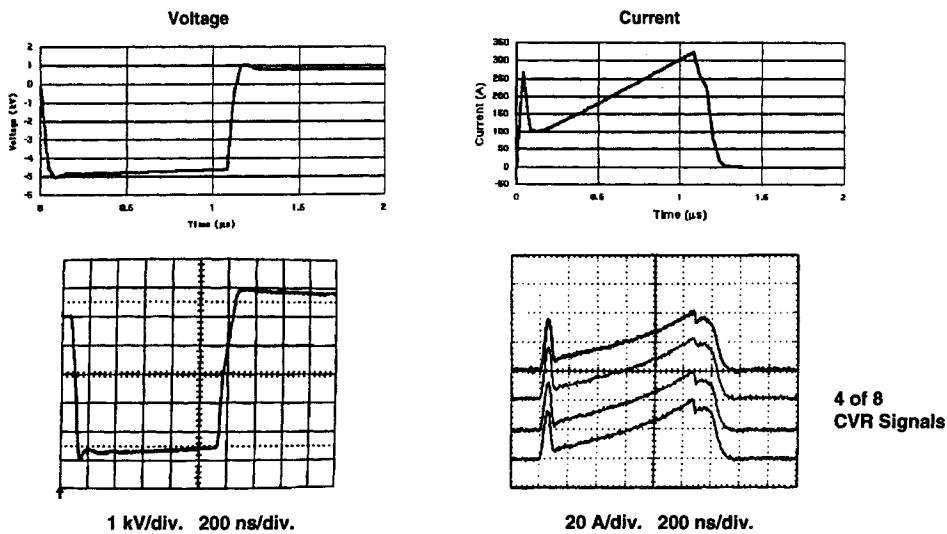


Figure 4. Comparison of a network simulation (top) with measured data (bottom).

shows that five, 200-ns pulses will recur at 250-kHz rate under a maximum prf performance test. Figure 5(b) illustrates the machine's pulse agility by altering the pulse width and pulse separation from pulse to pulse. For this test, the capacitor voltages remain at their maximum values but the pulse width changes from 200 ns for the first pulse to 1 μ s for the fifth pulse.

ARM Induction Modulator

Figure 6(a) is a photograph of the ARM induction modulator. An induction core is located behind the central plate and supported on an insulated horizontal shaft. Eight modules surround the core. Each module houses a capacitor bank inside a fiberglass outer shell and a switching package beneath the shell and against the core's outer surface. Figure 6(b) shows the detail of a single-switch package, where 24, 1-kV circuit cards are racked together and are interconnected by conducting washers.

The testing program for the ARM modulator began by characterizing the full machine at a low voltage (5 kV) to obtain network models and operating experience with various loads. Initial tests at high voltage were conducted with only two modules at a time to avoid unnecessary risk to the entire modulator. The final tests included eight modules at high voltage.

Figure 7 compares single-pulse voltage and current data with Micro-Cap IV simulations. The top row of simulated data was obtained for a 1- μ s pulse width with initial capacitor voltages of 5 kV and 1.5 kV. The bottom row of experimental data shows the acceleration voltage detail and the switch currents from two modules for the same initial conditions. In contrast to the recirculator, each module is equipped with one CVR, but eight current signals are still necessary to measure the total machine current.

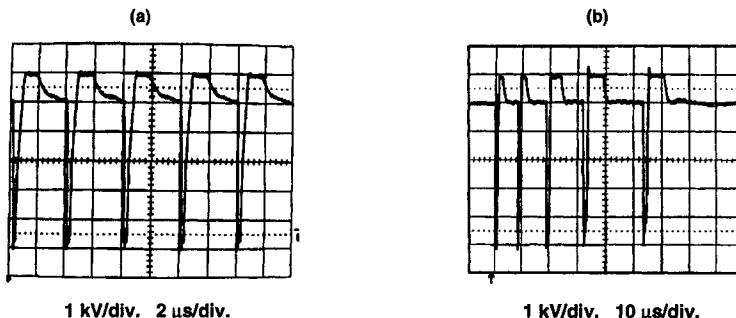
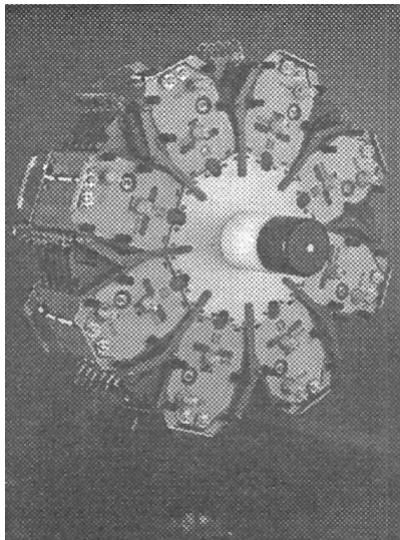


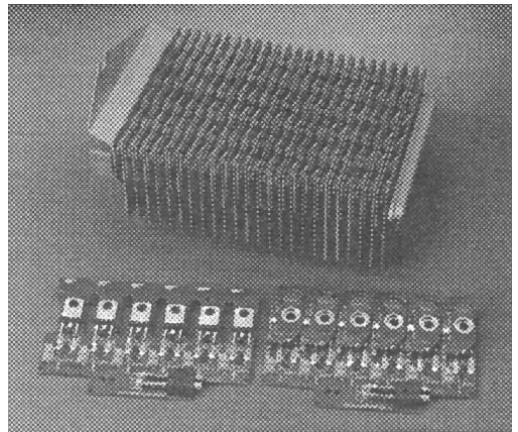
Figure 5. (a) Five-pulse burst of 200-ns pulses at a maximum rate of 250 kHz. (b) Five-pulse burst of variable pulses to show machine agility.

(a)



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(b)



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Figure 6. (a) Photograph of the ARM induction modulator. (b) Photograph of a single-switch package with two individual circuit cards.

Consequently, the total switch current depicted in the simulation can be compared with a measured value by multiplying either of the two current signals by eight.

Figure 8 shows the results of another single-pulse, 5-kV test, where we apply a long pulse until the induction core saturates. Figure 8(a) shows the core voltage collapse, which is matched by a sudden rise in switch current shown in Fig. 8(b). The flat portion of the switch current in Fig 8(b) indicates the current is limited by the FET array impedance to a total of 3 kA. The current falls to zero because the FETs are commanded to open at the end of the applied pulse.

Figure 9 shows two 5-pulse bursts of 200-ns pulses but under very different test conditions. Figure 9(a) is a 350-kHz burst from all eight modules charged to 5.3 kV. In this case, the modulator is driving a load impedance consisting of a 5- Ω resistive load in parallel with a 40- μ H inductor. The resistors simulate loading from an electron beam and are diode-coupled to conduct current only during the accelerating

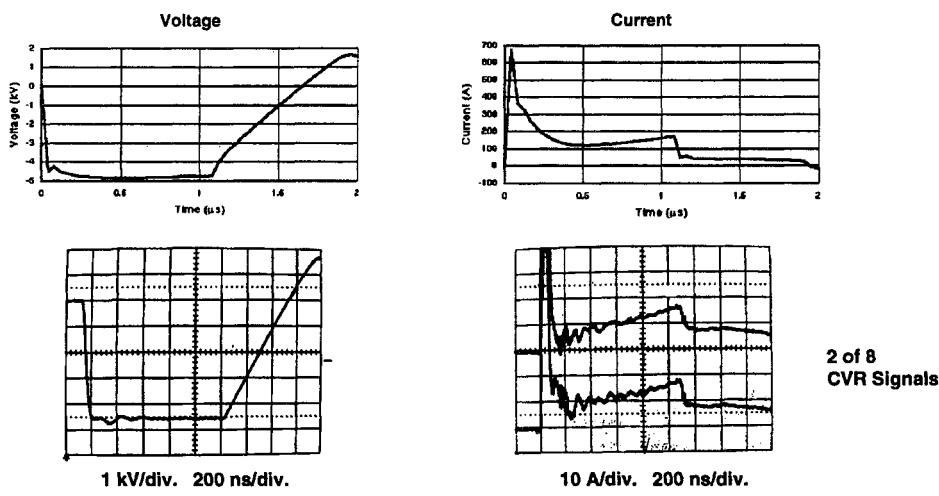


Figure 7. Comparison of a network simulation (top) with measured data (bottom).

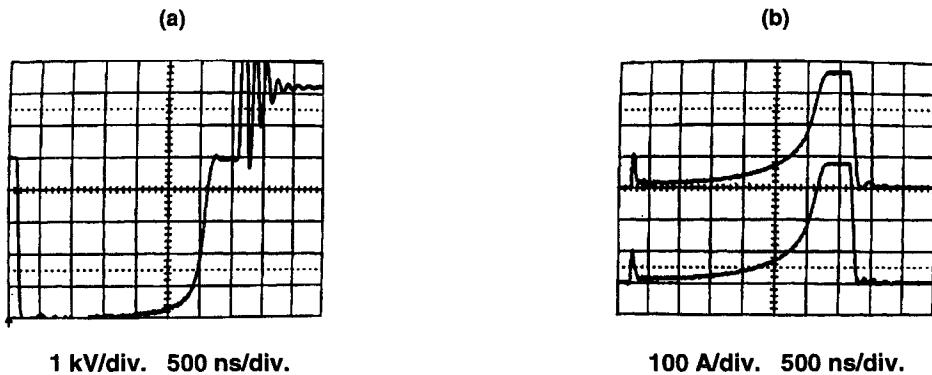


Figure 8. Response of the modulator to long pulses. (a) Core voltage and (b) switch current for two modules.

pulse. The 40- μ H inductor simulates the ARM accelerating cell and conducts current throughout the entire pulse. Each 200-ns pulse in Fig. 9(a) generates 1.1 kA of total switch current. Figure 9(b) is a 620-kHz burst from two modules charged to 14 kV. In this case, the modulator is unloaded so each 200-ns pulse generates only 75 A of total switch current. Other module pairs were tested to their full 15-kV working voltage.

Summary

To further enhance our understanding of accelerator designs, we tested solid-state power sources at LLNL. The prototypes are similar in their use of FET switching technology and basic circuit design; however, they differ in size and power. Test results indicate that agile, high-prf pulses can be generated by both prototypes while providing power to external loads for short bursts.

Acknowledgments

We wish to acknowledge the electronic and mechanical fabrication contributions by the EG&G personnel at the ATLAS facilities in Las Vegas, Nevada. In particular, the recent efforts of Bill Bellow and Ken Brose are appreciated. We also recognize the contributions by employees of Raytheon Corp. at the Nevada Test Site. Specifically, we thank Rob Saethre for his detailed analysis of our optical systems.

This work was performed under the auspices of the U.S. Department of Energy by the Lawrence Livermore National Laboratory under contract number W-7405-ENG-48.

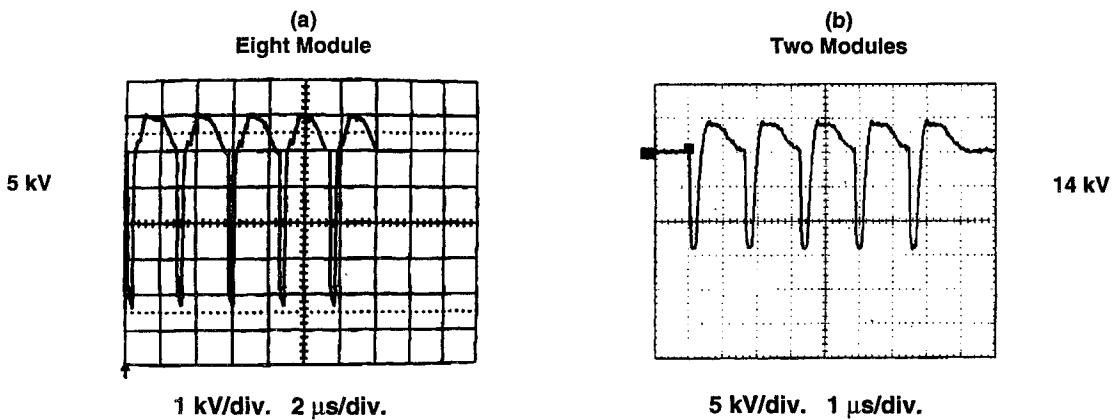


Figure 9. Five-pulse burst data for two test conditions. (a) Maximum prf test at 5 kV to 350 kHz. (b) Maximum prf test at 14 kV to 620 kHz.

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6. Micro-Cap IV is an electronic circuit analysis program published by Spectrum, 1021 South Wolfe Road, Sunnyvale, CA, 94086.